

Micro-programmed Control Unit(MCU)

PRODUCT DESCRIPTION

The MS616F512NS is a low dissipation 16-bit RISC MCU. It has five low power dissipation modes, which greatly increases battery lifespan of portable device . The digital oscillator (DCO) could arouse CPU from low dissipation mode within 6μs.



QFN48

FEATURES

- Low Power Supply : 1.8V-3.6V
- Ultra Low Dissipation
 - Operation Status : 280μA (1MHz,2.2V)
 - Standby Status: 1.1μA
 - Off Status (RAM hold): 0.1μA
- Five Power Saving Modes
- From Standby Mode to Wake-up Mode within 6μs
- 16-bit RISC Architecture, 125ns Instruction Period
- 16-bit Timer A with Three Capture/Compare Registers
- 16-bit Timer B with Seven Capture/Compare Registers
- A Integrated Comparator
- Serial Communication Interface(USART)
 - Optional Synchronous (UART) or Asynchronous(SPI) Mode
- Low Voltage Detection
- On-line Flash Programming via JTAG
- On-line Flash Programming via Bootstrap Loader
- 2KB RAM
- 62KB+128B Flash Memory

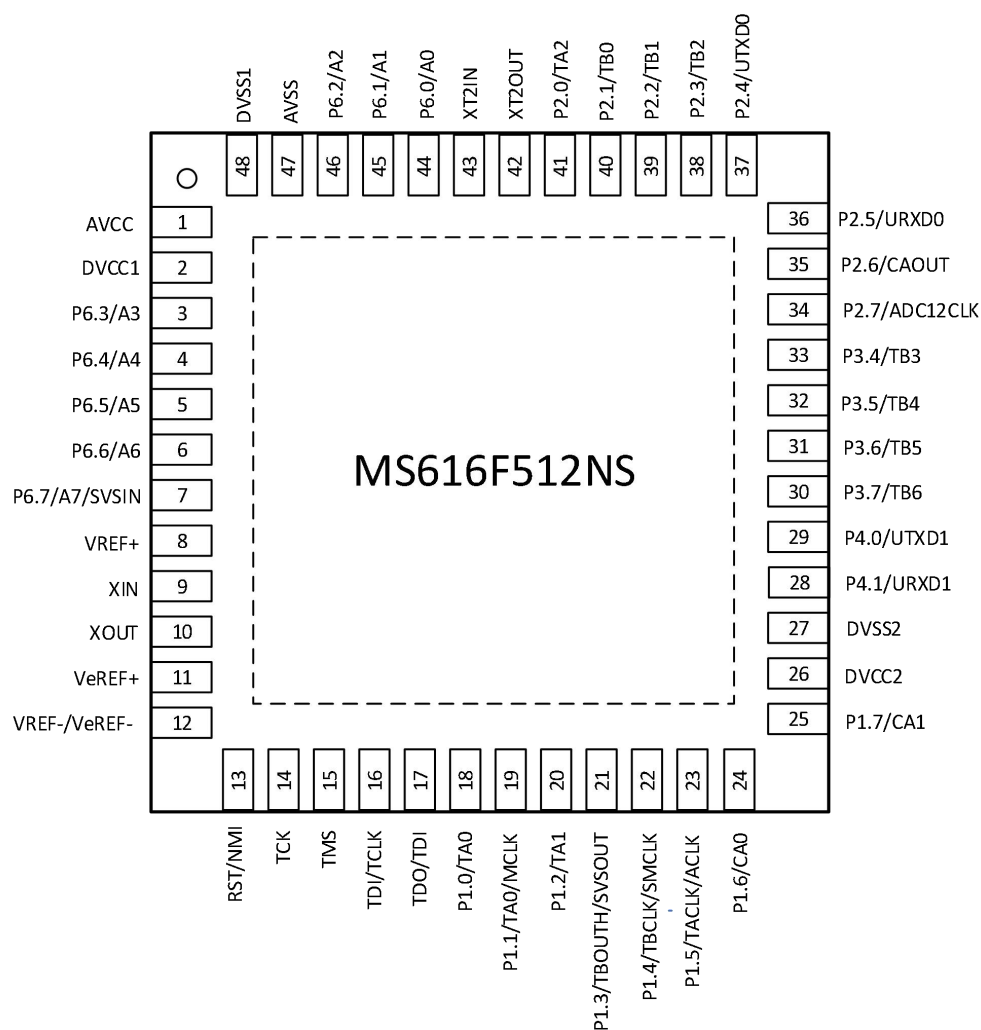
APPLICATIONS

- Measurement Area
- Industrial Control Area

PRODUCT SPECIFICATION

Part Number	Package	Marking
MS616F512NS	QFN48	MS616F512NS

PIN CONFIGURATION

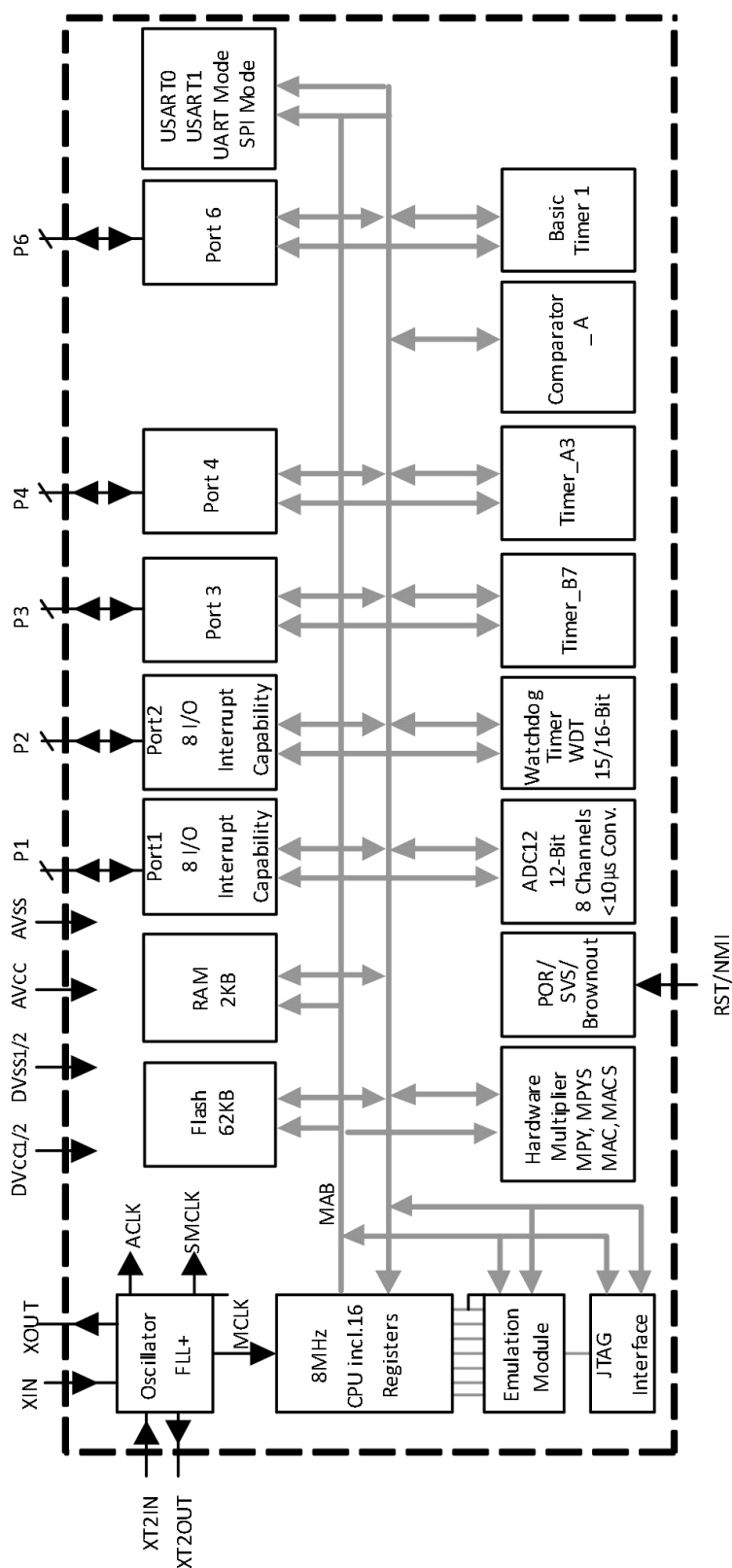


PIN DESCRIPTION

Pin	Name	Type	Description
1	AVCC	-	Analog Power Supply, Positive Terminal
2	DVCC1	-	Digital Power Supply, Positive Terminal
3	P6.3/A3	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A3
4	P6.4/A4	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A4
5	P6.5/A5	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A5
6	P6.6/A6	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A6
7	P6.7/A7/ SVSIN	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A7 / Analog Input of Supply Voltage Supervisor(SVS)
8	VREF+	O	ADC Internal Reference Voltage, Positive Output Terminal
9	XIN	I	Input Terminal of Crystal Oscillator XT1 , can connect to standard or clock crystal
10	XOUT	O	Output Terminal of Crystal Oscillator XT1
11	VeREF+	I	Input Terminal of ADC External Reference Voltage
12	VREF-/VeREF-	I	Negative Terminal of ADC Reference Voltage, including internal and external references
13	RST/NMI	I	Reset Input / No-shield Interrupt Input Terminal
14	TCK	I	Test Clock. TCK is clock input terminal for programming and test
15	TMS	I	Test Mode Selection. TMS is used as input terminal for programming and test
16	TDI/TCLK	I	Test Data or Clock Input. The protection fuse is connected to TDI/TCLK
17	TDO/TDI	I/O	Test Data Output. Data Output / Data Input
18	P1.0/TA0	I/O	General Digital I/O Interface / Capture Input of Timer_A : CCI0A , Compare Output: Out0 / BSL Output
19	P1.1/TA0/ MCLK	I/O	General Digital I/O Interface / Capture Input of Timer_A : CCI0B / Master Clock MCLK Output. TA0 (just as input pin)/ BSL Input
20	P1.2/TA1	I/O	General Digital I/O Interface / Capture Input of Timer_A : CCI1A , Compare Output: Out1
21	P1.3/TBOUTH/ SVSOUT	I/O	General Digital I/O Interface / All PWM Digital Output Terminals of Timer_B (TB0-TB6) Switched to High-impedance / SVS Comparator Output
22	P1.4/TBCLK/ SMCLK	I/O	General Digital I/O Interface / Timer_B Clock Signal, TBCLK Input / Submain System SMCLK Output
23	P1.5/TACLK/ ACLK	I/O	General Digital I/O Interface / Timer_A Clock Signal, TACLK Input / Auxiliary Clock ACLK Output (via 1, 2, 4, 8 frequency division)

Pin	Name	Type	Description
24	P1.6/CA0	I/O	General Digital I/O Interface / Comparator A Input
25	P1.7/CA1	I/O	General Digital I/O Interface / Comparator A Input
26	DVCC2	-	Digital Power Supply, Positive Terminal
27	DVSS2	-	Digital Power Supply, Negative Terminal
28	P4.1/URXD1	I/O	General Digital I/O Interface / In UART Mode, USART1 Receiving Data Input
29	P4.0/UTXD1	I/O	General Digital I/O Interface / In UART Mode, USART1 Transmitting Data Output
30	P3.7/TB6	I/O	General Digital I/O Interface / Capture Input of Timer_B7 CCR6 : CCI6A/CCI6B, Compare Output: Out6
31	P3.6/TB5	I/O	General Digital I/O Interface / Capture Input of Timer_B7 CCR5 : CCI5A/CCI5B, Compare Output: Out5
32	P3.5/TB4	I/O	General Digital I/O Interface / Capture Input of Timer_B7 CCR4 : CCI4A/CCI4B, Compare Output: Out4
33	P3.4/TB3	I/O	General Digital I/O Interface / Capture Input of Timer_B7 CCR3 : CCI3A/CCI3B, Compare Output: Out3
34	P2.7/ADC12CLK	I/O	General Digital I/O Interface / 12-bit ADC Conversion Clock
35	P2.6/CAOUT	I/O	General Digital I/O Interface / Comparator A Output
36	P2.5/URXD0	I/O	General Digital I/O Interface / In UART Mode, USART0 Receiving Data Input
37	P2.4/UTXD0	I/O	General Digital I/O Interface / In UART Mode, USART0 Transmitting Data Output
38	P2.3/TB2	I/O	General Digital I/O Interface / Capture Input of Timer_B7 CCR2 : CCI2A/CCI2B, Compare Output: Out2
39	P2.2/TB1	I/O	General Digital I/O Interface / Capture Input of Timer_B7 CCR1 : CCI1A/CCI1B, Compare Output: Out1
40	P2.1/TB0	I/O	General Digital I/O Interface / Capture Input of Timer_B7 CCR0 : CCI0A/CCI0B, Compare Output: Out0
41	P2.0/TA2	I/O	General Digital I/O Interface / Capture Input of Timer_A : CCI2A , Compare Output: Out2
42	XT2OUT	O	Output Terminal of Crystal Oscillator XT2
43	XT2IN	I	Input Terminal of Crystal Oscillator XT2
44	P6.0/A0	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A0
45	P6.1/A1	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A1
46	P6.2/A2	I/O	General Digital I/O Interface / 12-bit ADC Analog Input A2
47	AVSS	-	Analog Power Supply, Negative Terminal
48	DVSS1	-	Digital Power Supply, Negative Terminal

BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS

Any exceeding absolute maximum rating application causes permanent damage to device. Because long-time absolute operation state affects device reliability. Absolute ratings just conclude from a series of extreme tests. It doesn't represent chip can operate normally in these extreme conditions.

Parameter	Range	Unit
Voltage Difference From VCC to VSS	-0.3 ~ 4.1	V
Input Voltage	-0.3 ~ VDD+0.3	V
Device Diode Current	±2	mA
Operating Temperature (No Programming)	-55 ~ 150	°C
Storage Temperature (Programming)	-40 ~ 85	°C

Note: All voltages are relative to ground. V_{FB} , the JTAG fuse-blow voltage, could operate in maximum rating. When it needs to blow the fuse, TDI pin provides voltage.

RECOMMENDED OPERATING CONDITIONS

Parameter	Symbol	Condition	Min	Typ	Max	Unit
Power Supply	V _{CC}	Program executing	1.8		3.6	V
		Program executing and SVS enable, PORON=1	2.0		3.6	V
		Flash Programming	2.7		3.6	V
Power Supply	V _{SS}		0		0	V
Operating Temperature	T _A		-40		85	°C
LFXT1 Frequency	f _{LFXT1}	XTS_FLL=0, Quartz Oscillator		32.768		kHz
		XTS_FLL=1, Ceramic Oscillator	450		8000	kHz
		XTS_FLL=1, Crystal Oscillator	1000		8000	kHz
XT2 Frequency	f _{XT2}	Ceramic Oscillator	450		8000	kHz
		Crystal Oscillator	1000		8000	kHz
System Clock Frequency	F _{System}	V _{CC} =1.8V	DC		4.15	MHz
		V _{CC} =3.6V	DC		8	MHz

1. It's recommended to use the same power supply for AVCC and DVCC. The voltage difference between AVCC and DVCC can't exceed 0.3V.
2. When the power supply enough lows to trigger POR, the corresponding voltage is the minimum operating voltage. When the power supply increases to the value, which is equal to the minimum voltage value plus SVS hysteresis voltage, the POR signal stops.
3. In LF mode, the LFXT1 oscillator needs to connect with one external quartz oscillator. While in XT1 mode, LFXT1 needs to connect with one external ceramic or crystal oscillator.

ELECTRICAL CHARACTERISTICS

Supply Current

Parameter	Symbol	Condition	Min	Typ	Max	Unit
Operation Mode (Note 1) $f_{MCLK}=f_{SMCLK}=1\text{MHz}$, $f_{ACLK}=32768\text{Hz}$ $XTS_FLL=0, SELM=(0,1)$	I_{AM}	$T_A=-40^{\circ}\text{C}$ to 85°C	$V_{CC}=2.2\text{V}$	280	350	uA
			$V_{CC}=3\text{V}$	420	560	
Low Dissipation Mode (Note 1,4)	I_{LPM0}	$T_A=-40^{\circ}\text{C}$ to 85°C	$V_{CC}=2.2\text{V}$	32	45	uA
			$V_{CC}=3\text{V}$	55	70	
Low Dissipation Mode (Note 2,4) $f_{MCLK}=f_{SMCLK}=0\text{MHz}$ $f_{ACLK}=32768\text{Hz}, SCG=0$	I_{LPM2}	$T_A=-40^{\circ}\text{C}$ to 85°C	$V_{CC}=2.2\text{V}$	11	14	uA
			$V_{CC}=3\text{V}$	17	22	
Low Dissipation Mode (Note 2,4) $f_{MCLK}=f_{SMCLK}=0\text{MHz}$ $f_{ACLK}=32768\text{Hz}, SCG=1$	I_{LPM3}	$T_A=-40^{\circ}\text{C}$	$V_{CC}=2.2\text{V}$	1	1.5	uA
		$T_A=25^{\circ}\text{C}$		1.1	1.5	
		$T_A=-60^{\circ}\text{C}$		2	3	
		$T_A=85^{\circ}\text{C}$		3.5	6	
		$T_A=-40^{\circ}\text{C}$	$V_{CC}=3\text{V}$	1.8	2.2	
		$T_A=25^{\circ}\text{C}$		1.6	1.9	
		$T_A=-60^{\circ}\text{C}$		2.5	3.5	
		$T_A=85^{\circ}\text{C}$		4.2	7.5	
Low Dissipation Mode (Note 2,4) $f_{MCLK}=f_{SMCLK}=0\text{MHz}$ $f_{ACLK}=0\text{Hz}, SCG=1$	I_{LPM4}	$T_A=-40^{\circ}\text{C}$	$V_{CC}=2.2\text{V}$	0.1	0.5	uA
		$T_A=25^{\circ}\text{C}$		0.1	0.5	
		$T_A=-60^{\circ}\text{C}$		0.7	1.1	
		$T_A=85^{\circ}\text{C}$		1.7	3	
		$T_A=-40^{\circ}\text{C}$	$V_{CC}=3\text{V}$	0.1	0.5	
		$T_A=25^{\circ}\text{C}$		0.1	0.5	
		$T_A=-60^{\circ}\text{C}$		0.8	1.2	
		$T_A=85^{\circ}\text{C}$		1.9	3.5	

Note:

1. Timer_B frequency is locked as $f_{DCOCLK}=f_{DCO}=1\text{MHz}$. All inputs are connected to 0V or Vcc. All outputs have no source or reverse current.
2. All inputs are connected to 0V or Vcc. All outputs have no source or reverse current.
3. All inputs are connected to 0V or Vcc. All outputs have no source or reverse current. The current consumed of LPM3 is achieved through measuring the operating current of timer 1 and LCD (select ACLK). Comparator A and the current in SVS module would be described specially.

4. Including the current consumption in BROWNOUT module.

In operation mode, the relationship between consumption current and system frequency:

$$I_{(AM)} = I_{(AM)} [1 \text{ MHz}] \times f_{(\text{System})} [\text{MHz}]$$

In operation mode, the relationship between consumption current and power supply:

$$I_{(AM)} = I_{(AM)} [3 \text{ V}] + 175 \mu\text{A/V} \times (V_{CC} - 3 \text{ V})$$

Schmitt Trigger Input Terminal —P1, P2, P3, P4, P5, P6

Parameter		Power Supply	Min	Typ	Max	Unit
Forward Input Threshold Voltage	V_{IT+}	2.2 V	1.1		1.5	V
		3 V	1.5		1.9	
Reverse Input Threshold Voltage	V_{IT-}	2.2 V	0.4		0.9	V
		3 V	0.9		1.3	
Input Hysteresis ($V_{IT+} - V_{IT-}$)	V_{hys}	2.2 V	0.3		1.1	V
		3 V	0.5		1	

Standard Input Terminal —RST/NMI, JTAG(TCK, TMS, TDI, TDO)

Parameter		Power Supply	Min	Typ	Max	Unit
Low Level Input Voltage	V_{IL}	2.2V/3V	VSS		VSS + 0.6	V
High Level Input Voltage	V_{IH}		0.8VCC		VCC	V

Input Terminal—Px.x, TA_x, TB_x

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
External Interrupt Timing	$t_{(int)}$	Terminal P1, P2: P1.x to P2.x, External triggering signal is interrupt flag (Note 1)	2.2 V/3 V	1.5			cycle
			2.2 V	62			ns
			3 V	50			
Timer_A, Timer_B Capture Time	$t_{(cap)}$	TA0,TA1,TA2 TB0,TB1,TB2,TB3,TB4,TB5,TB6	2.2 V	62			ns
			3 V	62			
Clock Frequency applied to Timer_A/B	$f_{(TAext)}$	TACLK, TBCLK, INCLK: $t(H) = t(L)$	2.2 V			8	MHz
	$f_{(TBext)}$		3 V			10	
Clock Frequency of Timer_A/B	$f_{(TAint)}$	Select SMCLK or ACLK	2.2 V			8	MHz
	$f_{(TBint)}$		3 V			10	

Note 1: When external signal sets interrupt flag, the corresponding $t_{(int)}$ is even with trigger signal shorter than $t_{(int)}$. The clock cycle and time parameter must be simultaneously met to ensure interrupt flag setting. $t_{(int)}$ is measured referred to MCLK cycle.

Leakage Current (Note 1,2)

Parameter	Symbol	Condition	Power Supply	Min	Max	Unit
Px.x Terminal Leakage Current	I_{lkg}	Px Terminal: $V_{(Px.x)}$	2.2/3V		50	nA

Note :

1. Leakage current is measured when Vss or Vcc is applied to relative pins, unless otherwise noted.
2. The terminal pin must be set as input , and couldn't have any pull-up or pull-down resistor.

Output Terminal——P1, P2, P3, P4, P5, P6

Parameter	Condition	Power Supply	Min	Typ	Max	Unit
High Level Output Voltage	V_{OH}	$I_{OH(max)} = -1.5mA$ (Note 1)	2.2V	VCC-0.25	VCC	V
		$I_{OH(max)} = -6mA$ (Note 2)	2.2V	VCC-0.6	VCC	
		$I_{OH(max)} = -1.5mA$ (Note 1)	3V	VCC-0.25	VCC	
		$I_{OH(max)} = -6mA$ (Note 2)	3V	VCC-0.6	VCC	
Low Level Output Voltage	V_{OL}	$I_{OL(max)} = 1.5mA$ (Note 1)	2.2V	VSS	VSS+0.25	V
		$I_{OL(max)} = 6mA$ (Note 2)	2.2V	VSS	VSS+0.6	
		$I_{OL(max)} = 1.5mA$ (Note 1)	3V	VSS	VSS+0.25	
		$I_{OL(max)} = 6mA$ (Note 2)	3V	VSS	VSS+0.6	

Note :

1. $I_{OH(max)}$ and $I_{OL(max)}$ are the maximum total current, the sum of all output currents. Only when it is less than 12mA, just meet maximum voltage without drop.
2. $I_{OH(max)}$ and $I_{OL(max)}$ are the maximum total current, the sum of all output currents. Only when it is less than 48mA, just meet maximum voltage without drop.

Output Frequency

Parameter	Condition	Min	Typ	Max	Unit
$f(Px.y)$ ($1 \leq x \leq 6, 0 \leq y \leq 7$)	$C_L = 20 pF$, $I_L = 1.5 mA$	$V_{CC} = 2.2 V$	DC	5	MHz
		$V_{CC} = 3 V$	DC	7.5	
$f(ACLK)$	$C_L = 20 pF$			$f_{(System)}$	MHz
$f(MCLK)$					
$f(SMCLK)$					
Output Duty Cycle $t(Xdc)$	P1.5/TACLK/CLK, $C_L = 20 pF$ $V_{CC} = 2.2 V / 3 V$	$f_{(ACLK)} = f_{(LFXT1)} = f_{(XT1)}$	40%	60%	
		$f_{(ACLK)} = f_{(LFXT1)} = f_{(LF)}$	30%	70%	
		$f_{(ACLK)} = f_{(LFXT1)}$		50%	
	P1.1/TA0/MCLK, $C_L = 20 pF$ $V_{CC} = 2.2 V / 3 V$	$f_{(MCLK)} = f_{(XT1)}$	40%	60%	
		$f_{(MCLK)} = f_{(DCOCLK)}$	50%-15ns	50%	
	P1.4/TBCLK/SMCLK, $C_L = 20 pF$ $V_{CC} = 2.2 V / 3 V$	$f_{(SMCLK)} = f_{(XT2)}$	40%	60%	
		$f_{(SMCLK)} = f_{(DCOCLK)}$	50%-15ns	50%	

Wake-up Mode LPM3

Parameter	Condition	Power Supply	Min	Typ	Max	Unit
Delay Time	f = 1 MHz	2.2V/3V			6	us
	f = 2 MHz				6	
	f = 3 MHz				6	

RAM

Parameter	Condition	Min	Typ	Max	Unit
VRAMh	CPU Stop State (Note 1)	1.6			V

Note 1: The parameter defines the minimum power supply when RAM changes. And all programme must stop when measuring the parameter.

LCD

Parameter	Condition	Min	Typ	Max	Unit
Analog Voltage	V ₍₃₃₎ P5.7/R33 Voltage	2.5		VCC+0.2	V
	V ₍₂₃₎ P5.6/R23 Voltage		$[V_{(33)} - V_{(03)}] \times 2/3 + V_{(03)}$		V
	V ₍₁₃₎ P5.5/R13 Voltage		$[V_{(33)} - V_{(03)}] \times 2/3 + V_{(03)}$		V
	V ₍₃₃₎ -V ₍₀₃₎ R33 to R03 Voltage	2.2		VCC+0.2	V
Input Leakage Current	I _(R03) R03=VSS			20	nA
	I _(R13) P5.5/R13=VCC/3			20	nA
	I _(R23) P5.6/R23=2VCC/3			20	nA
Segment Address Voltage	V _(Sxx0)	V ₍₀₃₎		V ₍₀₃₎ -0.1	V
	V _(Sxx1)	V ₍₁₃₎		V ₍₁₃₎ -0.1	V
	V _(Sxx2)	V ₍₂₃₎		V ₍₂₃₎ -0.1	V
	V _(Sxx3)	V ₍₃₃₎		V ₍₃₃₎ -0.1	V

Comparator A (Note 1)

Parameter	Condition	Power Supply	Min	Typ	Max	Unit
I _(CC)	CAON=1, CARSEL=0, CAREF=0	2.2 V		25	40	uA
		3 V		45	60	
I _(Ref ladder/RefDiode)	CAON=1, CARSEL=0, No load at CAREF=1/2/3, P1.6/CA0 and P1.7/CA1	2.2 V		30	50	uA
		3 V		45	71	

Parameter	Condition	Power Supply	Min	Typ	Max	Unit
$V_{(Ref025)}$	PCA0=1, CARSEL=1, CAREF=1, No load at P1.6/CA0 and P1.7/CA1	2.2 V / 3 V	0.23	0.24	0.25	
$V_{(Ref050)}$	PCA0=1, CARSEL=1, CAREF=2, No load at P1.6/CA0 and P1.7/CA1	2.2V / 3 V	0.47	0.48	0.5	
$V_{(RefVT)}$	PCA0=1, CARSEL=1, CAREF=3, No load at P1.6/CA0 and P1.7/CA1; $T_A = 85^{\circ}\text{C}$	2.2 V	390	480	540	mV
		3 V	400	490	550	
Common-mode Input Voltage (V_{IC})	CAON=1	2.2 V / 3 V	0		VCC-1	V
Offset Voltage ($V_P - V_S$)	Note 2	2.2 V / 3 V	-30		30	mV
V_{hys}	CAON = 1	2.2 V / 3 V	0	0.7	1.4	mV
$t_{(response\ LH)}$	$T_A = 25^{\circ}\text{C}$, Overdrive 10mV, without filtering: CAF = 0	2.2 V	160	210	300	ns
		3 V	80	150	240	
	$T_A = 25^{\circ}\text{C}$, Overdrive 10mV, with filtering: CAF = 1	2.2 V	1.4	1.9	3.4	uA
		3 V	0.9	1.5	2.6	
$t_{(response\ HL)}$	$T_A = 25^{\circ}\text{C}$, Overdrive 10mV, without filtering : CAF = 0	2.2 V	130	210	300	ns
		3 V	80	150	240	
	$T_A = 25^{\circ}\text{C}$, Overdrive10mV, with filtering: CAF = 1	2.2 V	1.4	1.9	3.4	uA
		3 V	0.9	1.5	2.6	

Note:

1. The leakage current of comparator A has been defined in $Ilkg(Px.x)$.
2. By setting CAEX bit, make the comparator A input reverse. And measure twice continuously, the input offset voltage could be cancelled, then add the two measurements.

POR/Brownout Reset(BOR) (Note 1)

Parameter		Condition	Min	Typ	Max	Unit
td(BOR)	Brownout (Note 2)				2000	us
VCC(start)		dVCC/dt ≤ 3 V/s		0.7×V(B_IT-)		V
V(B_IT-)		dVCC/dt ≤ 3 V/s			1.71	V
Vhys(B_IT-)		dVCC/dt ≤ 3 V/s	70	130	180	mV
t(reset)		Pulse width needed at RST/NMI pin to reset internally, VCC = 2.2 V/3 V	2			us

Note:

1. The current consumed in Brownout module has been included in total currents I_{CC} . The voltage range is : $V(B_IT-) + V_{hys}(B_IT-) \leq 1.8V$.
2. After $V_{CC} = V(B_IT-) + V_{hys}(B_IT-)$, CPU starts to execute program after one $t_{d(BOR)}$ cycle. Before $V_{CC} \geq V_{CC(min)}$, FLL+ setting can't change. $V_{CC(min)}$ is the minimum power supply at operating frequency.

Supply Voltage Supervision(SVS)

Parameter	Condition		Min	Typ	Max	Unit
$t_{(SVSR)}$	$dV_{CC}/dt \geq 30 V/ms$		5		150	us
	$dV_{CC}/dt \leq 30 V/ms$				2000	us
$t_{d(SVSON)}$	SVSON, switch from VLD=0 to VLD \neq 0, $V_{CC} = 3 V$		20		150	us
t_{settle}	VLD \neq 0 (Note 2)				12	us
$V_{(SVSstart)}$	VLD \neq 0, $V_{CC}/dt \leq 3 V/s$			1.55	1.7	V
$V_{hys(SVS_IT-)}$	$V_{CC}/dt \leq 3 V/s$	VLD = 1	70	120	155	mV
		VLD = 2-14	$V_{(SVS_IT-)} \times 0.004$		$V_{(SVS_IT-)} \times 0.008$	
	$V_{CC}/dt \leq 3 V/s$ external voltage applied to A7 terminal	VLD = 15	4.4		10.4	mV
$V_{(SVS_IT-)}$	$V_{CC}/dt \leq 3 V/s$	VLD = 1	1.8	1.9	2.05	V
		VLD = 2	1.94	2.1	2.25	
		VLD = 3	2.05	2.2	2.37	
		VLD = 4	2.14	2.3	2.48	
		VLD = 5	2.24	2.4	2.6	
		VLD = 6	2.33	2.5	2.71	
		VLD = 7	2.46	2.65	2.86	
		VLD = 8	2.58	2.8	3	
		VLD = 9	2.69	2.9	3.13	
		VLD = 10	2.83	3.05	3.29	
		VLD = 11	2.94	3.2	3.42	
		VLD = 12	3.11	3.35	3.61(Note 1)	
		VLD = 13	3.24	3.5	3.76(Note 1)	
		VLD = 14	3.43	3.7(Note 1)	3.99(Note 1)	
	$V_{CC}/dt \leq 3 V/s$ external voltage applied to A7 terminal	VLD = 15	1.1	1.2	1.3	
$I_{CC(SVS)}$ (Note 3)	VLD \neq 0, $V_{CC} = 2.2 V/3 V$			10	15	uA

Note:

1. The maximum operating voltage is 3.6V.
2. t_{settle} is the settle time that comparator needs to stabilize level when VLD switches from not 0 to the value between 2 and 15. The overdrive voltage is assumed to be more than 50mV.
3. The consumption current of SVS module has been included in I_{cc}.

DCO

Parameter	Condition		Min	Typ	Max	Unit
f _(DCOCLK)	N(DCO)=01Eh, FN_8=FN_4=FN_3=FN_2=0,D=2; DCOPLUS = 1, f _{Crystal} =32.768kHz	VCC = 2.2 V/3V		1		MHz
f _(DCO=2)	FN_8=FN_4=FN_3=FN_2=0; DCOPLUS = 1	VCC = 2.2 V	0.3	0.65	1.25	MHz
		VCC = 3 V	0.3	0.7	1.3	MHz
f _(DCO=2)	FN_8=FN_4=FN_3=0, FN_2=1; DCOPLUS = 1	VCC = 2.2 V	0.7	1.3	2.3	MHz
		VCC = 3 V	0.8	1.5	2.5	MHz
f _(DCO=2)	FN_8=FN_4=0,FN_3=1,FN_2=x; DCOPLUS = 1	VCC = 2.2 V	1.2	2	3	MHz
		VCC = 3 V	1.3	2.2	3.5	MHz
S _n	S _n = f _{DCO(Tap n+1)} / f _{DCO(Tap n)}	1 < TAP ≤ 20	1.06		1.11	
		TAP = 27	1.07		1.17	
D _t	N(DCO) =01Eh, FN_8=FN_4=FN_3=FN_2=0, D= 2; DCOPLUS = 0	VCC = 2.2 V	-0.2	-0.3	-0.4	%/°C
		VCC = 3 V	-0.2	-0.3	-0.4	%/°C
D _v	N(DCO) =01Eh, FN_8=FN_4=FN_3=FN_2=0, D= 2; DCOPLUS = 0	VCC = 2.2 V/3V	0	5	15	%/V

Crystal Oscillator, LFXT1 Oscillator (Note 1, 2)

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Integrated Input Capacitance	C _{XIN}	OSCCAPx = 0h	2.2 V/3 V		0		pF
		OSCCAPx = 1h	2.2 V/3 V		10		
		OSCCAPx = 2h	2.2 V/3 V		14		
		OSCCAPx = 3h	2.2 V/3 V		18		
Integrated Output Capacitance	C _{XOUT}	OSCCAPx = 0h	2.2 V/3 V		0		pF
		OSCCAPx = 1h	2.2 V/3 V		10		
		OSCCAPx = 2h	2.2 V/3 V		14		
		OSCCAPx = 3h	2.2 V/3 V		18		
Input Logic on XIN	V _{IL}	Note 3	2.2 V/3 V	VSS		0.2VCC	V
	V _{IH}			0.8VCC		VCC	V

Note:

1. The parasitic capacitance generated from package and board is about 2pF. The crystal effective load capacitance is $(C_{XIN} \times C_{XOUT}) / (C_{XIN} + C_{XOUT})$, which has nothing with XTS_FLL.
2. There are some principles to be observed as follows, in order to improve the EMI characteristic of low power dissipation LFX1 oscillator, especially in LF mode(32kHz).
 - (1) The traces between the MS616F512NS and crystal should be as short as possible.
 - (2) Optimal design of ground plane near oscillator pin.
 - (3) Avoid that other clock and data lines have crosstalk with XIN and XOUT pins.
 - (4) Avoid layout traces below or near XIN and XOUT pins.
 - (5) By using match materials and repeated practices to reduce parasitic load on XIN, XOUT pins.
 - (6) If use protection coating, pay attention not to cause capacitive and resistive leakage among oscillator pins.
3. Only valid when using external logic clock and must set XTS_FLL bit. While invalid when using crystal and resonator.
4. For accurate real time clock application, OSCCAPx=0h, apply recommended capacitance.

Crystal Oscillator, XT2 Oscillator (Note 1)

Parameter		Condition	Min	Typ	Max	Unit
Integrated Input Capacitance	C _{XT2IN}	V _{CC} = 2.2 V/3 V		2		pF
Integrated Output Capacitance	C _{XT2OUT}	V _{CC} = 2.2 V/3 V		2		pF
Input Logic on XT2IN	V _{IL}	V _{CC} = 2.2 V/3 V	V _{SS}		0.2V _{CC}	V
	V _{IH}	Note 2	0.8V _{CC}		V _{CC}	V

Note: 1. The two terminals of oscillator all need to connect with load capacitance. The accurate capacitance value is provided by crystal manufacturer.

2. Only valid when using external logic clock and must set XTS_FLL bit. While invalid when using crystal and resonator.

USART0, USART1 (Note 1)

Parameter		Condition	Min	Typ	Max	Unit
USART0/1:	t _(τ)	V _{CC} = 2.2 V, SYNC = 0, UART mode	200	430	800	ns
Deglitch Time		V _{CC} = 3 V, SYNC = 0, UART mode	150	280	500	

Note 1: The signal applied to USART0/1 receiving terminal should meet the t_(τ) timing requirement, thus ensure URXS trigger is set. URXS trigger is set by low level pulse, which satisfies t_(τ) minimum timing requirement. The operating conditions set by flag bit must be independent of the timing constraint. The deglitch circuit only operates when URXD0/1 line transmit negatively.

12-bit ADC, Power Supply and Input Range (Note 1)

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Analog Power Supply	AVCC	AVCC and DVCC is connected together, AVSS and DVSS is connected together, $V_{(AVSS)} = V_{(DVSS)} = 0\text{ V}$		2.2		3.6	V
Analog Input Voltage (Note 2)	$V_{(P6.x/Ax)}$	Applicable to all P6.0/A0 to P6.7/A7 terminals. Analog input terminal is selected by ADC12MCTLx and P6Sel.x=1, $0 \leq x \leq 7$; $V_{(AVSS)} \leq V_{P6.x/Ax} \leq V_{(AVCC)}$		0		V_{AVCC}	V
Operating Current on AVCC (Note 3)	I_{ADC12}	$f_{ADC12CLK} = 5.0\text{ MHz}$ ADC12ON = 1, REFON = 0 SHT0=0, SHT1=0, ADC12DIV=0	2.2 V		0.65	1.3	mA
			3 V		0.8	1.6	
Operating Current on AVCC (Note 4)	I_{REF+}	$f_{ADC12CLK} = 5.0\text{ MHz}$ ADC12ON = 0, REFON = 1, REF2_5V = 1	3 V		0.5	0.8	mA
		$f_{ADC12CLK} = 5.0\text{ MHz}$ ADC12ON = 0, REFON = 1, REF2_5V = 0	2.2 V		0.5	0.8	
			3 V		0.5	0.8	mA
Input Capacitance	C_i	Choose only one terminal, P6.x/Ax	2.2 V			40	pF
Input Multiplexer Resistance	R_i	$0\text{ V} \leq V_{Ax} \leq V_{AVCC}$	3 V			2000	Ω

Note :

1. Leakage current has been defined in P.x/Ax terminal parameter sheet.
2. Analog input voltage range should be within reference voltage range, thus achieve valid conversion result.
3. Reference current is not included in I_{ADC12} .
4. Reference current is provided by AVCC. And the current is independent of ADC12ON until conversion starts.
Before A/D conversion, set REFON bit to enable built-in reference voltage module.

12-bit ADC, External Reference

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Positive External Reference Voltage Input	V_{eREF+}	$V_{eREF+} > V_{REF-}/V_{eREF-}$ (Note 2)		1.4		V_{AVCC}	V
Negative External Reference Voltage Input	V_{REF-}/V_{eREF-}	$V_{eREF+} > V_{REF-}/V_{eREF-}$ (Note 3)		0		1.2	V
External Reference Differential Voltage Input	$(V_{eREF+} - V_{REF-}/V_{eREF-})$	$V_{eREF+} > V_{REF-}/V_{eREF-}$ (Note 4)		1.4		V_{AVCC}	V
Static Input Current	I_{VeREF+}	$0\text{ V} \leq V_{eREF+} \leq V_{AVCC}$	2.2 V/3 V			± 1	μA
Static Input Current	$I_{VREF-/VeREF-}$	$0\text{ V} \leq V_{eREF+} \leq V_{AVCC}$	2.2 V/3 V			± 1	μA

Note :

1. External reference charges and discharge capacitance array during conversion. For external reference, input capacitance C_i is dynamic load during conversion period. The dynamic impedance of reference voltage should be matched with analog source impedance recommendation, achieving 12-bit setup accuracy on charging.
2. The accuracy constrains the minimum of positive external reference voltage. Decreasing accuracy demand could use lower reference voltage.
3. The accuracy constrains the maximum of negative external reference voltage. Decreasing accuracy demand could use higher reference voltage.
4. The accuracy constrains the difference of external reference voltage. Decreasing accuracy demand could use lower difference reference voltage.

12-bit ADC, Built-in Reference

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Positive Built-in Reference Voltage	V_{REF+}	REF2_5V = 1, $I_{VREF+} \leq I_{VREF+max}$	3V	2.4	2.5	2.6	V
		REF2_5V = 0, $I_{VREF+} \leq I_{VREF+max}$	2.2 V/3 V	1.44	1.5	1.56	
Minimum Power Supply of Positive Built-in Reference Voltage	$AVCC_{(min)}$	REF2_5V = 0, $I_{VREF+} \leq 1mA$		2.2			V
		REF2_5V = 1, $I_{VREF+} \leq 0.5mA$		$V_{REF+} + 0.15$			
		REF2_5V = 1, $I_{VREF+} \leq 1mA$		$V_{REF+} + 0.15$			
Load Current on V_{REF+}	I_{VREF+}		2.2 V	0.01		-0.5	mA
			3V			-1	
Load Modulation Current on V_{REF+}	$I_{L(VREF)+}$	$I_{VREF+} = 500 \mu A \pm 100 \mu A$ Analog input voltage $\sim 0.75V$; REF2_5V = 0	2.2 V			± 2	LSB
			3V			± 2	LSB
		$I_{VREF+} = 500 \mu A \pm 100 \mu A$ Analog input voltage $\sim 1.25 V$; REF2_5V = 1	3V			± 2	LSB
Load Modulation Time on V_{REF+}	$I_{DL(VREF)+}$	$I_{VREF+} = 100 \mu A \rightarrow 900 \mu A$, $C_{VREF+} = 5 \mu F$, Analog input voltage $\sim 0.5V_{REF+}$ Conversion result error ≤ 1 LSB	3V			20	ns
External Capacitance on V_{REF+} (Note 1)	C_{VREF+}	REFON = 1, $0 mA \leq I_{VREF+} \leq I_{VREF+max}$	2.2 V/3 V	5	10		μF
Temperature Coefficient of Built-in Reference	T_{REF+}	I_{VREF+} is a constant, range: $0mA \leq I_{VREF+} \leq 1mA$	2.2 V/3 V			± 100	ppm/ $^{\circ}C$
Setup Time of Built-in Reference (Note 2)	t_{REFON}	$I_{VREF+} = 0.5mA$, $C_{VREF+} = 10\mu F$, $V_{REF+} = 1.5 V$	2.2 V			17	ms

Note :

1. Internal buffer magnifier and accuracy demand need one external capacitor. For all INL and DNL tests, connect two capacitors between VREF+ and AVSS, VREF-/VREF- and AVSS, a 10uF tantalum capacitor and a 100nF ceramic capacitor.
2. Test condition: the conversion error is less than ± 0.5 LSB after t_{REFON} opens. The setup time is up to external capacitance load.

12-bit ADC, Timing Parameter

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
$f_{ADC12CLK}$		Ensure accuracy of ADC linearity parameter	2.2V/ 3 V	0.45	5	6.3	MHz
Internal ADC12 Oscillator	$f_{ADC12OSC}$	ADC12DIV=0, $f_{ADC12CLK}=f_{ADC12OSC}$	2.2 V/ 3 V	3.7		6.3	MHz
Conversion Time	$t_{CONVERT}$	$C_{VREF+} \geq 5$ uF, internal oscillator, $f_{ADC12OSC} = 3.7$ MHz - 6.3 MHz	2.2V/ 3 V	2.06		3.51	us
		External $f_{ADC12CLK}$ from ACLK, MCLK or SMCLK: ADC12SSEL $\neq 0$			$13 \cdot ADC12DIV \cdot 1/f_{ADC12CLK}$		us
ADC Enable Time	$t_{ADC12ON}$	Note 1				100	ns
Sample Time	t_{Sample}	$R_S = 400 \Omega$, $R_I = 1000 \Omega$, $C_I = 30$ pF, $\tau = [R_S + R_I] \times C_I$; (Note 2)	3 V	1220			ns
			2.2 V	1400			

Note :

1. After ADC12ON is enabled, $t_{ADC12ON}$ is the time when the conversion error is less than ± 0.5 LSB. And the reference voltage and input signal have be set.
2. After about 10τ , the conversion error is less than ± 0.5 LSB, $t_{Sample} = \ln(2n+1) \times (R_S + R_I) \times C_I + 800ns$ ($n=ADC$ resolution=12, R_S = input resistance).

12-bit ADC, Linearity Parameter

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Integral Nonlinearity Error	E_I	$1.4 V \leq (V_{eREF+} - V_{REF-}/V_{eREF-}) \min \leq 1.6 V$	2.2 V /3 V			± 2	LSB
		$1.6 V < (V_{eREF+} - V_{REF-}/V_{eREF-}) \min \leq [V_{(AVCC)}]$				± 1.7	
Differential Nonlinearity Error	E_D	$(V_{eREF+} - V_{REF-}/V_{eREF-}) \min \leq (V_{eREF+} - V_{REF-}/V_{eREF-})$, $C_{VREF+} = 10$ uF (tantalum) and 100 nF (ceramic)	2.2 V/ 3 V			± 1	LSB
Offset Error	E_O	$(V_{eREF+} - V_{REF-}/V_{eREF-}) \min \leq (V_{eREF+} - V_{REF-}/V_{eREF-})$, Internal impedance of source $R_S < 100\Omega$, $C_{VREF+} = 10$ uF (tantalum) and 100 nF (ceramic)	2.2 V/ 3 V		± 2	± 4	LSB

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Gain Error	E_G	$(V_{eREF+} - V_{REF-}/V_{eREF-}) \min \leq (V_{eREF+} - V_{REF-}/V_{eREF-})$, $C_{VREF+} = 10 \mu F$ (tantalum) and 100 nF (ceramic)	2.2 V / 3 V		± 1.1	± 2	LSB
Total Unadjusted Error	E_T	$(V_{eREF+} - V_{REF-}/V_{eREF-}) \min \leq (V_{eREF+} - V_{REF-}/V_{eREF-})$, $C_{VREF+} = 10 \mu F$ (tantalum) and 100 nF (ceramic)	2.2 V / 3 V		± 2	± 5	LSB

12-bit ADC, Temperature Sensor and Built-in Mid-voltage V_{MID}

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Current on AV_{CC} (Note 1)	I_{SENSOR}	REFON = 0, INCH = 0Ah, ADC12ON=NA, $T_A = 25$	2.2 V		40	120	uA
			3 V		60	160	
V_{SENSOR}		ADC12ON =1, INCH = 0Ah, $T_A = 0^\circ C$	2.2 V		986	986 $\pm 5\%$	mV
			3 V		986	986 $\pm 5\%$	
TC_{SENSOR}		ADC12ON = 1, INCH = 0Ah	2.2 V		3.55	3.55 $\pm 3\%$	mV/ $^\circ C$
			3 V		3.55	3.55 $\pm 3\%$	
Sample Time Required by Selecting Channel 10 (Note 2)	t_{sensor} (sample)	ADC12ON = 1, INCH = 0Ah, Conversion error ≤ 1 LSB	2.2 V	30			us
			3 V	30			
Current Divided on Channel 11	I_{VMID}	ADC12ON = 1, INCH = 0Bh (Note 3)	2.2 V			NA	uA
			3 V			NA	
Voltage Divided on Channel 11	V_{MID}	ADC12ON = 1, INCH = 0Bh, V_{MID} is $\sim 0.5 \times V_{AVCC}$	2.2 V		1.1	1.1 ± 0.04	V
			3 V		1.5	1.5 ± 0.04	
Sample Time Required by Selecting Channel 11	t_{VMID} (sample)	ADC12ON = 1, INCH = 0Bh, Conversion error ≤ 1 LSB	2.2 V	1400			ns
			3 V	1220			

Note :

1. If (ADC12ON=1, REFON=1) or (ADC12ON=1, INCH=0Ah and sample signal at high-level), sensor current I_{SENSOR} would be generated. The current includes the values through sensor and reference.
2. Typical equivalent impedance of sensor is 51k Ω . Sample time includes sensor enabled time $t_{SENSOR(on)}$.
3. V_{MID} is only used during sampling process, without generating extra current.
4. The sample time, $t_{VMID(sample)}$ has included enabled time $t_{VMID(on)}$, without extra time.

Flash

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
Operating Voltage at Programming, Erasing	VCC (PGM/ERASE)			2.7		3.6	V
The Frequency for Flash Programming Timing	fFTGP			257		476	kHz
The Frequency for Flash Erasing Timing	fFTGE			15		100	kHz
Current on DVCC at Programming	I _{PGM}		2.7 V/ 3.6 V		3	5	mA
Current on DVCC at Erasing	I _{ERASE}		2.7 V/ 3.6 V		3	7	mA
Accumulated Programming Time	t _{CPT}	Note 1	2.7 V/ 3.6 V			10	ms
Accumulated Large-scale Erasing Time	t _{CMErase}	Note 2	2.7 V/ 3.6 V	200			ms
Programming/ Erasing Duration				10 ⁴	10 ⁵		cycles
Data Save Cycle	t _{Retention}	T _J = 25°C		100			years
Word Programming Time	t _{Word}	Note 3			35		t _{FTG}
Block Programming Time of the First Word	t _{Block, 0}				30		
Block Programming Time of Each Additional Word	t _{Block, 1-63}				21		
Waiting Time for Block Programming Finishing Sequencing	t _{Block, End}				6		
Large-scale Erasing Time	t _{Mass Erase}				5297		
Segment Erasing Time	t _{Seg Erase}				4819		

Note:

1. When a 64-bit Flash module is written, it couldn't exceed accumulated programming time. The parameter is applicable to all Flash programming methods.
2. The large-scale erasing time is up to Flash timing.
At least 11.1ms (= 5297x1/f_{FTG}, maximum = 5297x1/476kHz).
3. These values have been fixed into the state machine of Flash controller.
4. The erasing frequency on chip is less than 100K.
5. Program on chip by word format, rather than byte format.
6. Information storage only has A segment (128 bytes), without B segment.
7. 2048 bytes rather than 512 bytes in each segment of master storage.

JTAG, Interface

Parameter		Condition	Power Supply	Min	Typ	Max	Unit
TCK Input Frequency	f _{TCK}	Note 1	2.2 V	0		5	MHz
			3 V	0		10	MHz
Internal Pull-up Resistors on TMS, TCK, TDI/TCLK	R _{internal}	Note 2	2.2 V/ 3 V	25	60	90	kΩ

Note :

1. f_{TCK} may be constrained by the timing requirement of selected module.
2. TMS, TDI/TCLK and TCK pull-up resistors have been integrated.

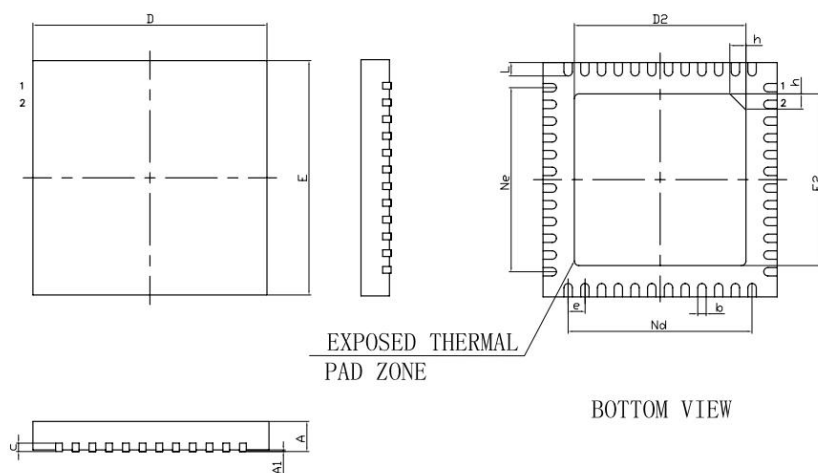
JTAG, Fuse (Note 1)

Parameter		Condition	Min	Typ	Max	Unit
Power Supply Needed by Fuse-blow	$V_{CC(FB)}$	$T_A = 25^{\circ}\text{C}$	2.5			V
TDI/TCLK Voltage Needed by Fuse-blow	V_{FB}		6		7	V
TDI/TCLK Current Needed by Fuse-blow	I_{FB}				100	mA
Time for Fuse-blow	t_{FB}				1	ms

Note 1: Once the fuse is blown, the JTAG/Test of the MS616F512NS can't be connected, and the simulation characteristic would be lost. JTAG mode is switched to bypass mode.

PACKAGE OUTLINE DIMENSIONS

QFN48



Symbol	Dimensions in Millimeters		
	Min	Typ	Max
A	0.70	0.75	0.80
A1	-	0.02	0.05
b	0.15	0.20	0.25
c	0.18	0.20	0.23
D	5.90	6.00	6.10
D2	4.10	4.20	4.30
e	0.40BSC		
Ne	4.40BSC		
Nd	4.40BSC		
E	5.90	6.00	6.10
E2	4.10	4.20	4.30
L	0.35	0.40	0.45
h	0.30	0.35	0.40
L/F Size (mil)	177×177		

MARKING and PACKAGING SPECIFICATIONS

1. Marking Drawing Description



Product Name : MS616F512NS

Product Code : XXXXXXXX

2. Marking Drawing Demand

Laser printing, contents in the middle, font type Arial.

3. Packaging Specifications

Device	Package	Piece/Reel	Reel/Box	Piece /Box	Box/Carton	Piece/Carton
MS616F512NS	QFN48	2000	1	2000	8	16000

STATEMENT

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- The process of improving product is endless. And our company would sincerely provide more excellent product for customer.

**MOS CIRCUIT OPERATION PRECAUTIONS**

Static electricity can be generated in many places. The following precautions can be taken to effectively prevent the damage of MOS circuit caused by electrostatic discharge:

1. The operator shall ground through the anti-static wristband.
2. The equipment shell must be grounded.
3. The tools used in the assembly process must be grounded.
4. Must use conductor packaging or anti-static materials packaging or transportation.



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